

EE.351: Spectrum Analysis and Discrete-Time Systems

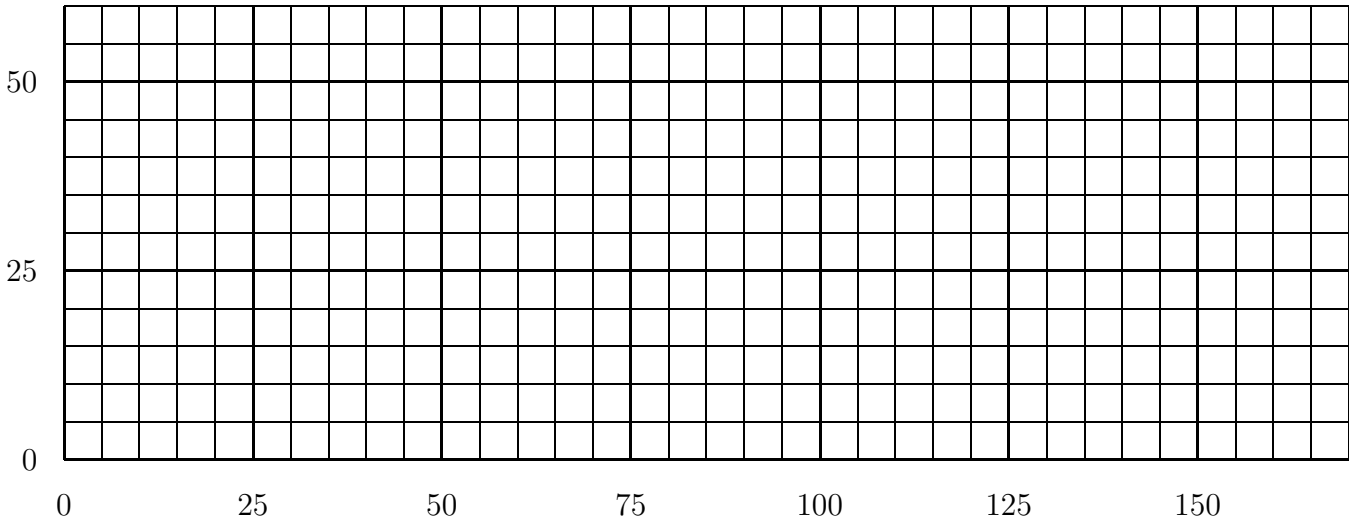
FINAL EXAM, 9:00AM–12:00PM, December 12, 2003 (closed book)

Examiner: Ha H. Nguyen

*Note:* There are six questions. All questions are of equal value but not necessarily of equal difficulty. Full marks shall only be given to solutions that are properly explained and justified.

1. (*FS of a continuous-time signal*) Consider a half-wave rectifier whose output  $y(t)$  is related to the input  $x(t)$  as follows:  $y(t) = \begin{cases} x(t), & \text{if } x(t) \geq 0 \\ 0, & \text{if } x(t) < 0 \end{cases}$ . Let  $x(t) = \cos(2\pi t)$ .

[3] (a) Neatly sketch the input  $x(t)$  and the output  $y(t)$ . What are the fundamental periods of  $x(t)$  and  $y(t)$ ?



[5] (b) Determine the Fourier series coefficients of the output  $y(t)$ .

[2] (c) What are the amplitudes of the dc components of the input and output signals, respectively?

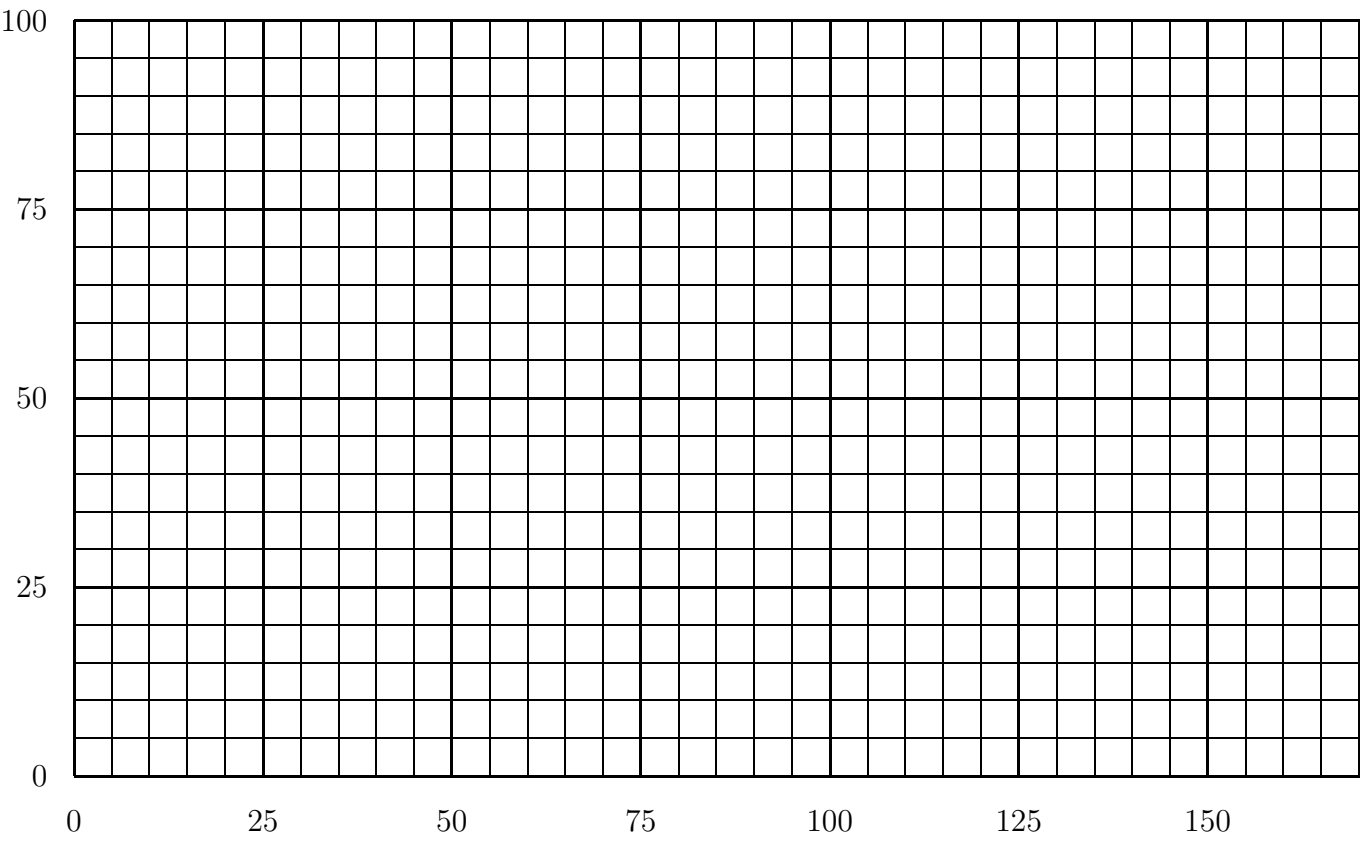
2. (*FS of a discrete-time signal*) Consider the following discrete-time periodic signal

$$x[n] = 2 + 2 \cos\left(\frac{\pi}{3}n - \frac{\pi}{4}\right) + \sin\left(\frac{2\pi}{3}n - \frac{\pi}{6}\right)$$

- [2] (a) What are the fundamental frequency and fundamental period of  $x[n]$ ?

- [5] (b) Find the Fourier series coefficients for  $x[n]$ .

[3] (c) Plot the magnitude and phase spectrum of  $x[n]$  over two periods.

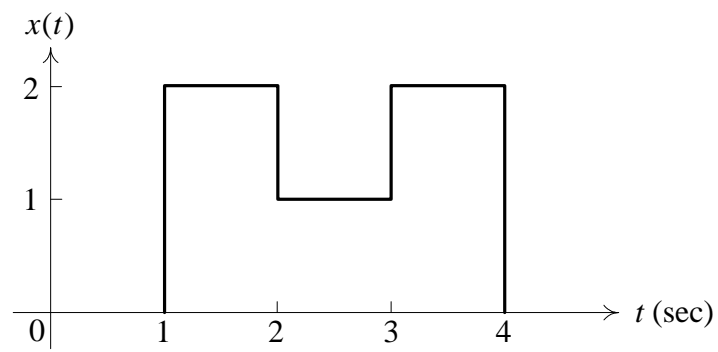


3. (*Fourier Transform*) The signal  $x(t)$  has Fourier transform  $X(j\omega)$ .

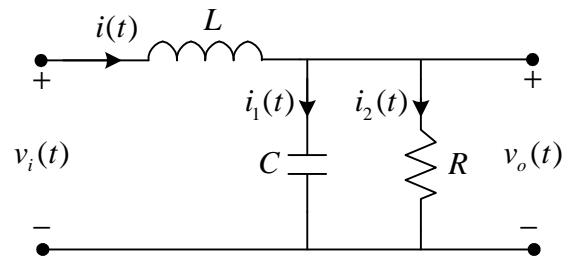
[2] (a) Show that the Fourier transform of  $dx(t)/dt$  is  $(j\omega)X(j\omega)$ .

[2] (b) Show that the Fourier transform of  $x(t - t_0)$  is  $e^{-j\omega t_0}X(j\omega)$ , where  $t_0$  is a constant.

[6] (c) Find  $X(j\omega)$  if  $x(t)$  is given below.



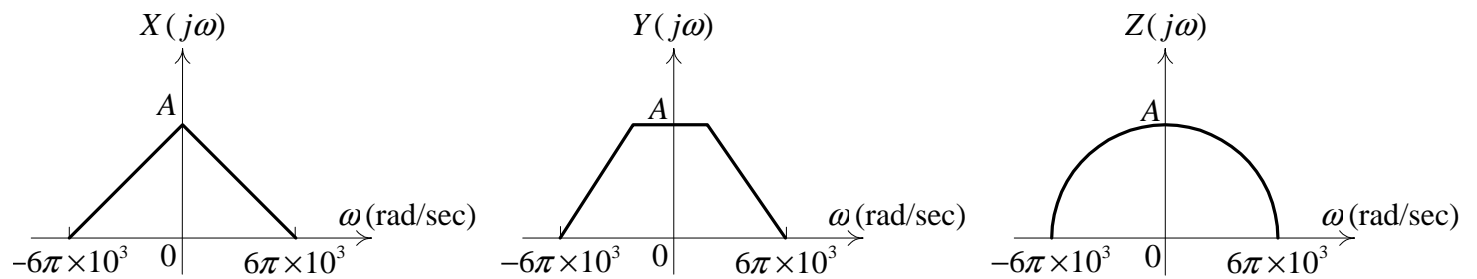
4. (Frequency response of an RLC filter) Consider the following circuit:



- [5] (a) Show that the frequency response of the circuit is given by  $H(j\omega) = \frac{1}{1 - \omega^2 LC + j\omega L/R}$

- [5] (b) Let  $\omega_c = \sqrt{\frac{1}{LC}}$  and  $L = 2R^2C$ . Show that the magnitude frequency response of the circuit can be written as  $|H(j\omega)| = \frac{1}{\sqrt{1 + (\omega/\omega_c)^4}}$ . Is the circuit a low-pass or high-pass filter? Explain your answer.

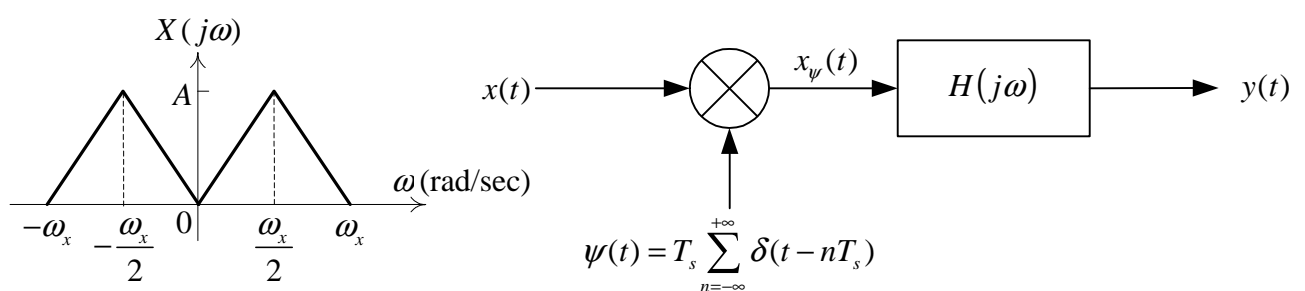
5. (*Amplitude modulation*) The manager of your division asked you to design a communication system to transmit three signals  $x(t)$ ,  $y(t)$  and  $z(t)$  simultaneously over a radio channel. The spectra of three signals are shown below. You bought a radio frequency spectrum from 650 kHz to 668 kHz (i.e., from  $2\pi \times 650 \times 10^3$  rad/sec to  $2\pi \times 668 \times 10^3$  rad/sec) and decided to use amplitude modulation.



- [5] (a) Clearly present a block diagram of the transmitter to your manager. Also show him the spectrum of the transmitted signal. Clearly identify all the relevant frequencies.

- [5] (b) To convince your manager that your design is valid, show and explain to him a block diagram of the receiver that can perfectly recover the signal  $y(t)$  (assume that there is no distortion introduced by the channel).

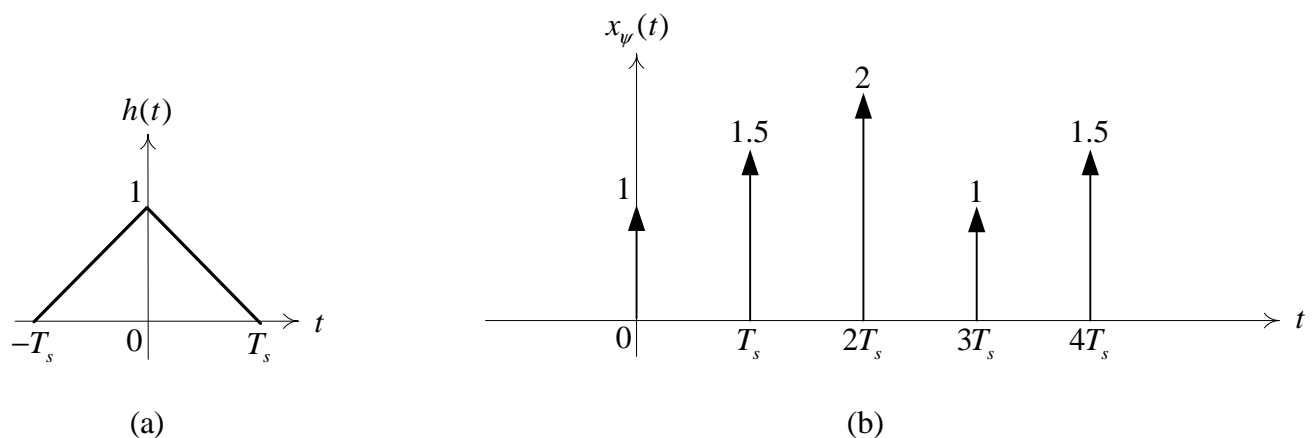
6. (*Sampling*) A block diagram of impulse sampling is shown below.



- [2] (a) If the sampling frequency is  $\omega_s = \frac{2\pi}{T_s} = \frac{3\omega_x}{2}$ , does aliasing occur? What is the minimum sampling frequency to prevent aliasing?

- [5] (b) Draw the spectrum of  $x_\psi(t)$  when  $\omega_s = \frac{3\omega_x}{2}$ . Also draw the spectrum of  $y(t)$  if  $H(j\omega)$  is an ideal low-pass filter with a cutoff frequency of  $\omega_c = \omega_x$ .

- [3] (c) Instead of the ideal low-pass filter, consider the filter in Figure (a) as a reconstruction filter. Find and plot the reconstructed signal  $y(t)$  if  $x_\psi(t)$  is as shown in Figure (b) over the interval  $[0, 4T_s]$ .





**Potentially Useful Facts:**

- FS representation of DT periodic signals (DTFS):

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 n}, \quad a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk\omega_0 n}$$

- FS representation of CT periodic signals (CTFS):

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}, \quad a_k = \frac{1}{T_0} \int_{T_0} x(t) e^{-jk\omega_0 t} dt$$

- FT of CT signals:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega, \quad X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt$$

- Fourier transform pairs/properties:

$$\begin{aligned} \cos(\omega_0 t) &\xleftrightarrow{\mathcal{FT}} \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0) \\ \delta(t) &\xleftrightarrow{\mathcal{FT}} 1 \\ x(t - t_0) &\xleftrightarrow{\mathcal{FT}} e^{-j\omega t_0} X(j\omega) \\ \frac{d}{dt}x(t) &\xleftrightarrow{\mathcal{FT}} j\omega X(j\omega) \\ \int_{-\infty}^t x(\tau) d\tau &\xleftrightarrow{\mathcal{FT}} \frac{1}{j\omega} X(j\omega) \quad (\text{Ignoring DC signal}) \\ \cos(\omega_0 t)x(t) &\xleftrightarrow{\mathcal{FT}} \frac{1}{2}X(j(\omega - \omega_0)) + \frac{1}{2}X(j(\omega + \omega_0)) \\ x(t)\psi(t) &\xleftrightarrow{\mathcal{FT}} \sum_{k=-\infty}^{\infty} X(j(\omega - k\omega_s)) \quad (\text{Impulse sampling}) \end{aligned}$$

- Identities:

$$\begin{aligned} e^{j\omega n} &= \cos(\omega n) + j \sin(\omega n) \\ e^{-j\omega n} &= \cos(\omega n) - j \sin(\omega n) \\ \cos(x) \cos(y) &= \frac{1}{2}[\cos(x + y) + \cos(x - y)] \end{aligned}$$

For my own survey, please indicate (✓) which equations provided above are useful for you, i.e., the ones that you used and you might not remember if not provided. Thanks.